Performance Study of Hybrid DS/FFH Spread-Spectrum Systems in the Presence of Multipath Fading and Multiple-Access Interference

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Outline

- Introduction
- Types of SS Systems
- Multipath Fading Channel
- Direct Digital Synthesizers (DDS)
- Performance Evaluation of DS/FFH
- Numerical Results
- Conclusion



ORNL's Solution: Hybrid Spread Spectrum

- Novel technique (3 Patents); adaptive, programmable.
- HSS: a synergistic combination of DS, FH, and TH.
- Advantages:
 - Adaptive Hybrid Spread-Spectrum (HSS) modulation format combines DSSS and frequency/time hopping in a multidimensional, orthogonal signaling scheme.
 - Capable of excellent LPI, LPD & security properties (programmable).
 - > Adaptive, robust protocol for high QoS applications.
 - Can be operated in burst mode for very low power drain.
 - Superior resistance to multipath and jamming (high process gain).
 - Easily deployed with modern chip technology.
 - Compliant with existing FCC/NTIA/ETSI rules for ISM bands.
 - Ideally implemented via modern FPGA-based electronics, ASICs, and SDR techniques.

Typical HSS Applications



DIRECT-SEQUENCE SPREAD-SPECTRUM



spreading with PN sequence

(PN Rate >> Data Rate)

same as and lined up with received PN sequence



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before spreading

FREQUENCY-HOPPING SPREAD-SPECTRUM



Narrow spectrum at output of modulator while hopping Spectrum has same bandwidth and power density after hopping with PN sequence (PN Rate << Data Rate for standard FHSS) Original narrowband, high power data stream is restored if local PN sequence is same as and lined up with received PN sequence



SLOW HYBRID SPREAD-SPECTRUM (DS/SFH)



DS (wide) spectrum at output of modulator while hopping

Spectrum has same bandwidth and power density after hopping with PN sequence (PN Rate << Data Rate for standard HSS; >> for FastHSS™) Original narrowband, high power data stream is restored if local PN sequence is same as and lined up with received PN sequence



FAST HYBRID SPREAD-SPECTRUM (DS/FFH)



 $G_{p(FH/DS) dB} = G_{p(FH) dB} + G_{p(DS) dB} = 10 \log (no. of hopping channels) + 10 \log (BW_{DS}/R_{info})$



HSS is a Multidimensional Signal

- HSS can be defined in 3 axes (code, frequency, and time).
 - Each dimension is orthogonal with the others.
 - Permissible signal spaces along an axis may also be ~ orthogonal.

 $(C_0, F_0, t_0)^{T}$

 (C_4, F_2, t_0)

Code

 (C_1, F_2, t_2)

 (C_3, F_0, t_1)

 (C_2, F_1, t_0)

- Codes
- Frequencies
- Time slots
- Easily adaptable to exploit

many degrees of freedom to

meet system requirements.

Some signal overlaps may be orthogonal. Time

MULTIPATH PROPAGATION



ORNL is developing new techniques to mitigate short-path degradations!



Direct Digital Synthesizers (DDS)



- The implementation of a DDS has two distinct parts:
 - A phase accumulator accumulates the phase increment and adds in the phase offset.
 - The DDS output is then calculated by quantizing the results of the phase accumulator section and using them to select values from a lookup table.



RF Coexistence – Multi-User, Jamming, and Multipath Fading Performance

The error probability of one hop is:

- K Number of users
- M Number of FH channels
- *L* Number of hops per bit
- $T = \frac{T_b}{L}$ Duration of each hop

 $T_c = \frac{T_b}{NL}$ Chip duration for *PN* sequence

- N Period of the PN sequence
- $W_{\rm DS}$ Bandwidth of DS waveform
- W_{J} Bandwidth of the wideband jamming
- W Number of hopping channels corrupted by jamming
- W_J^p Part of the bandwidth of the channel partially corrupted by jamming

$$\begin{split} P_{h} &= \sum_{j=0}^{K-1} P_{\varepsilon}^{k}(j \text{ users}) \\ &= \sum_{j=0}^{K-1} \{P_{\varepsilon}^{k}(j \text{ users, no } jam) + P_{\varepsilon}^{k}(j \text{ users, ful } jam) + P_{\varepsilon}^{k}(j \text{ users, partial } jam)\} \\ &= \sum_{j=0}^{K-1} \{P(j \text{ users, no } jam)P^{k}(\varepsilon \mid j \text{ users, no } jam) \\ &+ P(j \text{ users, ful } jam)P^{k}(\varepsilon \mid j \text{ users, ful } jam) \\ &+ P(j \text{ users, partial } jam)P^{k}(\varepsilon \mid j \text{ users, partial } jam)\} \end{split}$$

The error probability of one bit is:

$$P_{b} = \sum_{d=\frac{L+1}{2}}^{L} \binom{L}{d} (P_{h})^{d} (1-P_{h})^{L-d}$$



RF Coexistence – Multi-User, Jamming, and Multipath Fading Performance

- *P* Transmitted signal power
- M Number of FH channels
- *L* Number of hops per bit

$$T = \frac{T_b}{L}$$
 Duration of each hop

- $T_c = \frac{T_b}{NL}$ Chip duration for *PN* sequence
 - N Period of the PN sequence
 - W_{DS} Bandwidth of DS waveform
 - W_{J} Bandwidth of the wideband jamming
 - W Number of hopping channels corrupted by jamming
 - W_J^p Part of the bandwidth of the channel partially corrupted by jamming

 $NSR = N_0 / 2PT$ Noise-to-signal ratio

 $JSR = N_J / 2PT$ Jamming-to-signal ratio

 I_j^{κ} Interference to signal ratio introduced by the other users hopping in user k's channel

 $q = W_J^p / W_{DS}$ Fraction of the channel jammed

Partial Jamming

$P^{k}(\varepsilon \mid j \text{ users, partial jam}) = q P^{k}(\varepsilon \mid j \text{ users, partial jam, corrupted portion}) + (1-q)P^{k}(\varepsilon \mid j \text{ users, partial jam, uncorrupted portion})$

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From the problem formulation, we have

$$P(j \text{ users, no } jam) = {\binom{K-1}{j}} \left(\frac{1}{M}\right)^{j} \left(1 - \frac{1}{M}\right)^{K-1-j} \left(\frac{M-W-1}{M}\right)$$
$$P(j \text{ users, full } jam) = {\binom{K-1}{j}} \left(\frac{1}{M}\right)^{j} \left(1 - \frac{1}{M}\right)^{K-1-j} \left(\frac{W}{M}\right)$$
$$P(j \text{ users, partial } jam) = {\binom{K-1}{j}} \left(\frac{1}{M}\right)^{j} \left(1 - \frac{1}{M}\right)^{K-1-j} \left(\frac{1}{M}\right)$$

$$P^{k}(\varepsilon \mid j \text{ users, no } jam) = Q\left(\frac{1}{\sqrt{NSR/2 + I_{j}^{k}}}\right)$$

Full Jamming

$$P^{k}(\varepsilon \mid j \text{ users, full jam}) = Q\left(\frac{1}{\sqrt{NSR/2 + JSR/2 + I_{j}^{k}}}\right)$$



- Severe (~worst-case) channel conditions:
 - K = 20-120 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance $\lambda = 10$
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough



Performance of a hybrid DS/FFH system: Effect of different number of users (multi-user interference).



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 10-16 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance $\lambda = 10$
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
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Performance of a hybrid DS/FFH system: Effect of different jamming-tonoise ratios (JNRs).



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance $\lambda = 1$
 - Jammed chan. = 2-8
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
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Performance of a hybrid DS/FFH system: Effect of different number of fully-jammed channels.



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 1-7 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance $\lambda = 1$
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
 ¹⁷ Managed by UT-Battelle folgeneration Energy



Performance of a hybrid DS/FFH system: Effect of different numbers of frequency hops per bit.



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 10-40 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance $\lambda = 10$
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
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Performance of a hybrid DS/FFH system: Effect of different numbers of available hopping channels.



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 4-64 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance λ = 10
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
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Performance of a hybrid DS/FFH system: Effect of different DS PN code lengths.



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1-0.7
 - Chan. covariance $\lambda = 10$
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
 ²⁰ Managed by UT-Battelle folgeneration Energy



Performance of a hybrid DS/FFH system: Effect of different Rician fading channel parameters.



- Severe (~worst-case) channel conditions:
 - K = 100 users, 100% duty cycle
 - L = 5 hops/bit
 - M = 20 channels
 - N = 127 PN code length
 - DS PG = 21 dB
 - JNR = 13 dB
 - Rician coeff. γ = 0.1
 - Chan. covariance λ = 10
 - Jammed chan. = 5
 - Channel portion partially corrupted = 0.4
- Represents user flooding in rough
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Performance of a hybrid DS/FFH system: Effect of different hitting rates.



Hybrid DS/FFH Comparative Simulation

- Severe (~worst-case) channel conditions:
 - Two-path Rayleigh chan.
 - P1: ∆ = 0, gain = 0.7
 - P2: Δ = 0.3 μs, gain = 0.4
 - Equal-bandwidth cases
 - DS SF = 16
 - FH: 16 ch., 4 b/hop
 - DS/FH: SF = 16, 4 hopping freqs.
- Represents the fixedbandwidth advantage of DS/FFH format over other modulations.



Comparative performance of a hybrid DS/FFH versus other forms in Rayleigh fading.



Conclusions

- The performance of a hybrid DS/FFH system was analytically evaluated in a worst-case use scenario.
- We derived the average BER for a hybrid DS/FFH system that includes the effects from wide-band and partial-band jamming, multi-user interference and/or varying degrees of Rician/Rayleigh fading.
- Numerical results exploring the parameter space of the HSS system have been presented to demonstrate its effectiveness under different conditions and scenarios.
- The detailed performance and security aspects of HSS signals will be further analyzed in a future paper.





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